# Maintaining Impedance in Split Core Ferrites for Low-Frequency Applications

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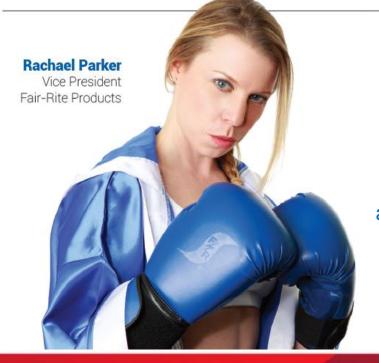


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Joining Fair-Rite in 2014, she holds a BS in Electrical Engineering, an M.Eng in Electrical Engineering, and an MS in Engineering Management. Rachael spent her early career in product development, project leadership, and program management.





# Fair-Rite Products Corp.

#### **Your Signal Solution®**

- Founded in 1952 family owned and operated in US and China
- Providing ferrite components for the electronics industry for over 60 years
- Wide product offering and materials for:
  - EMI Suppression
  - RFID-Antennas
  - Power/Inductive
  - Value Added Services (machining, winding, assemblies)
- Visit our new website at www.fair-rite.com





## The challenge...

- So you've made it to the test lab to get your product certified – but you fail!!
  - The dreaded questions:
    - What is the fix?
    - How long will it take?
    - How much will it cost??



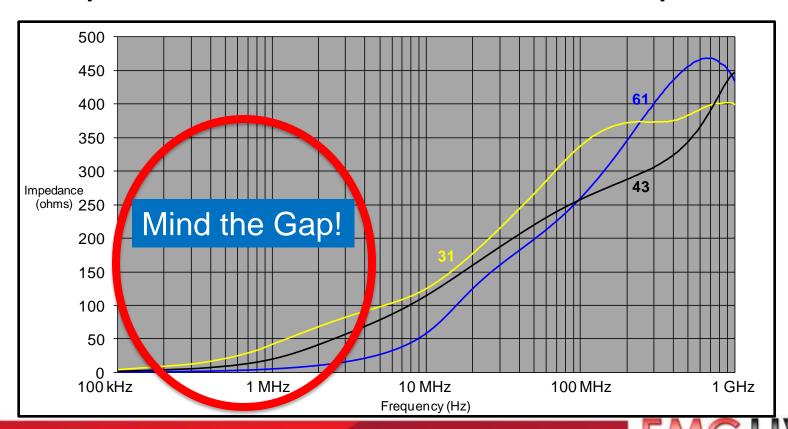




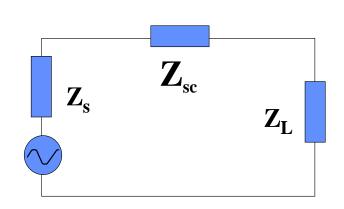


#### Until now...

To solve low-frequency EMC issues,
 Snap-It ferrites were not a viable option



#### Review: How Ferrites Reduce Noise

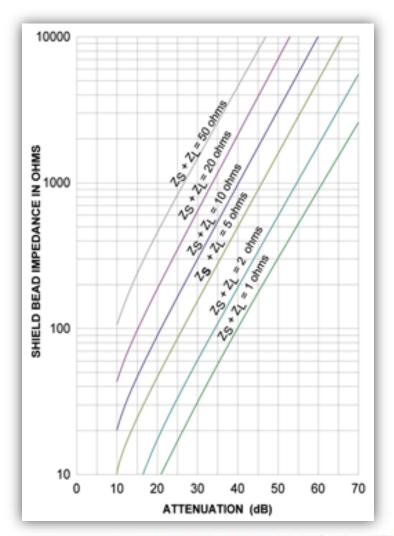


$$Attenuation = 20 \cdot \log_{10} \frac{Z_S + Z_{SC} + Z_L}{Z_S + Z_L}$$

Z<sub>s</sub> = Source impedance

Z<sub>sc</sub> = Suppressor Core impedance

= Load impedance

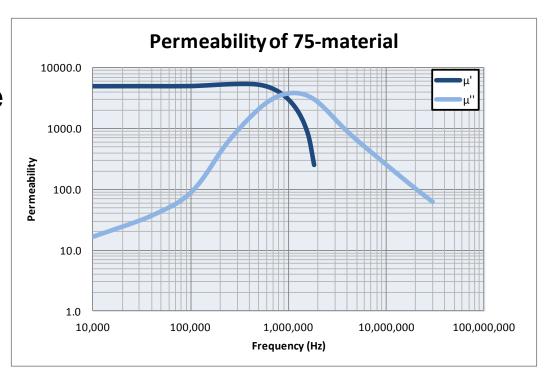




$$Z = j\omega L_s + R_s = j\omega L_o(\mu' - j\mu'') \quad [\Omega]$$

 $L_s$  = series inductance  $R_s$  = series loss resistance  $L_o$  = air core inductance  $\mu$  = material permeability

Material permeability is a **frequency dependant** characteristic





$$Z = j\omega L_s + R_s = j\omega L_o(\mu' - j\mu'') \quad [\Omega]$$

 Generally, as the initial permeability (μ<sub>i</sub>) increases, the impedance frequency range decreases

Material	μ <sub>i</sub> (initial permeability)	Frequency Range
75	5000	150 kHz – 10 MHz
31	1500	1 MHz – 300 MHz
43/44	800/500	25 MHz – 300 MHz
61	125	200 MHz – 1 GHz

Initial permeability refers to  $\mu$  when magnetization levels extremely small



$$Z = j\omega L_s + R_s = j\omega \frac{L_o(\mu' - j\mu'')}{[\Omega]}$$

 $L_s$  = series inductance

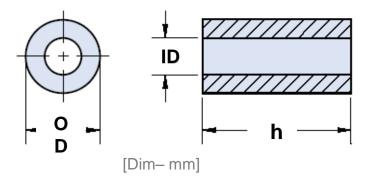
 $R_s$  = series loss resistance

 $L_0$  = air core inductance

 $\mu$  = material permeability

$$L_o = \frac{\mu_0 N^2 A_e}{l_e} = \frac{\mu_0 N^2 \cdot h \cdot \ln\left(\frac{OD}{ID}\right)}{2\pi}$$
 [H]

Ae = effective cross-sectional area  $l_e$  = effective magnetic path length  $\mu_0$  = permeability of free space N = number of turns





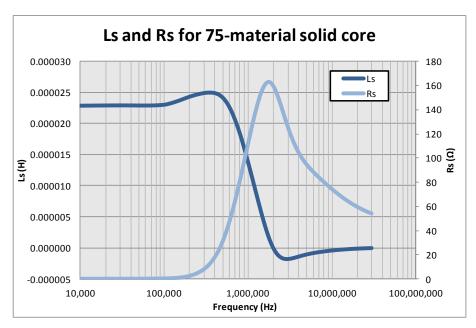
$$Z = j\omega L_s + R_s = j\omega \frac{L_o(\mu' - j\mu'')}{L_o(\mu' - j\mu'')} \quad [\Omega]$$

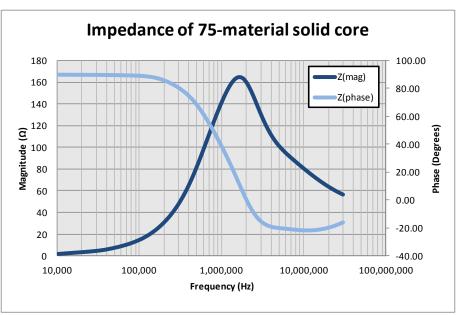
 Core performance also depends on geometric and implementation parameters as part of L<sub>o</sub>:

Parameter	Description	As parameter is increased, Z will
N	Number of times the cable passes through aperture the core	Increase by N <sup>2</sup>
OD	Outer Diameter of the core	Increase linearly
ID	Inner Diameter of the core	Decrease linearly
Н	Height of the core	Increase linearly



## Putting it all together...

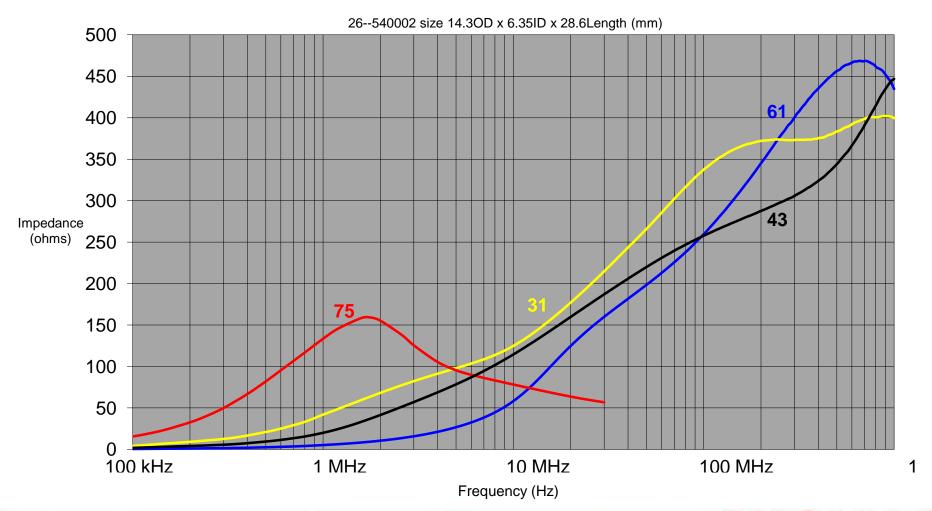




- Low frequencies → L<sub>s</sub> is main driver of impedance.
- High frequencies (> 1MHz)→ R<sub>s</sub> is main driver of impedance



## **Solid Suppression Core Comparison**





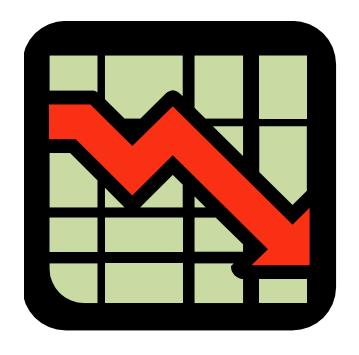
#### Challenges



Test facilities had implementation difficulties with solid cores.



#### Challenges

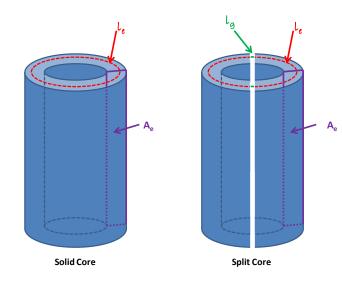


Performance was degraded in split cores due the high permeability material (e.g.,  $\mu_i$  = 5000)



## Impedance with an Air-Gap

- Air-gap in the magnetic circuit reduces the inductance.
  - $l_{\rm e}$  needs to be considered separately.
- Core behaves as if it has reduced permeability, known as its effective permeability.
  - Permeability is dependant on both the magnetic path length and the air gap.





## Impedance with an Air-Gap

Assuming gap length (I<sub>g</sub>) << magnetic path lengh (I<sub>e</sub>)

$$Z = j\omega \left(\frac{\mu_o N^2 h \ln\left(\frac{r_o}{r_i}\right)}{2\pi}\right) \left(\frac{\mu_e'}{-\frac{j\mu_e''}}\right) \left[\Omega\right]$$

$$tan(\delta)_{gapped} = \frac{\tan(\delta)}{\mu'} \mu'_e$$

$$\mu'_e = \frac{l_e}{\frac{l_e}{\mu'} + l_g}$$

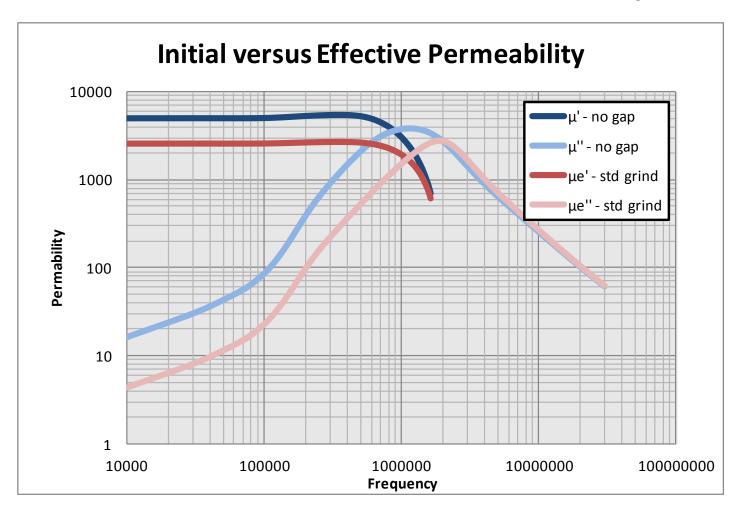
$$tan(\delta) = \frac{\mu''}{\mu'}$$

 $\mu''_e = \mu'_e \tan(\delta)_{\text{gapped}}$ 

tan(δ) refers to the magnetic loss tangent and represents the inefficiency of the system based on the angle between the μ' and μ".



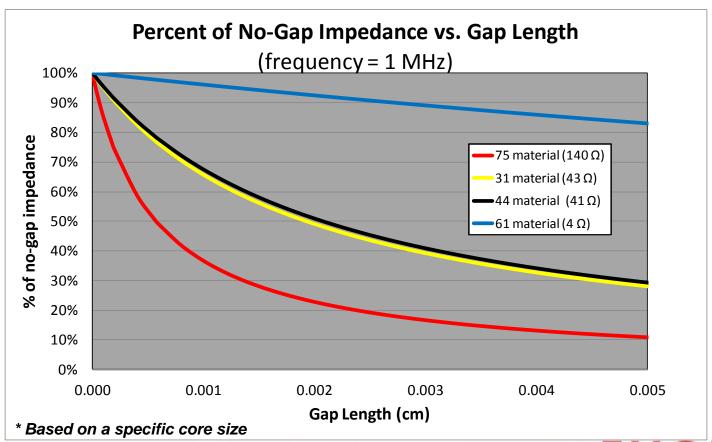
## Effect on Permeability





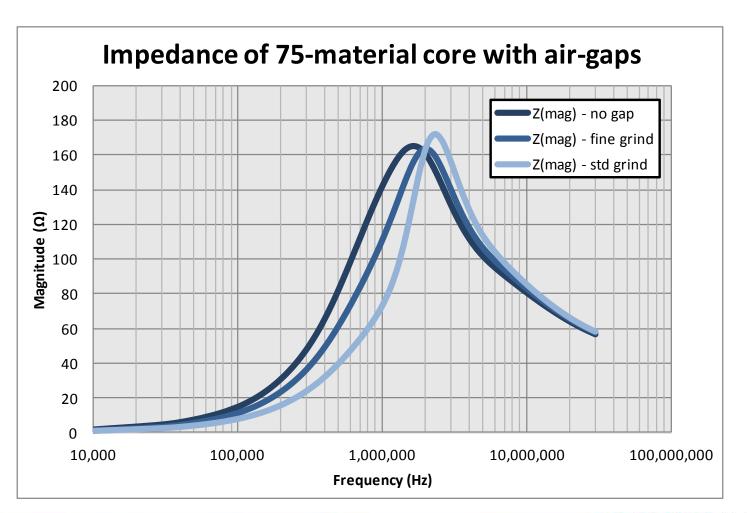
#### So what??

 For a high-permeability such as 75-material, even the smallest gaps can degrade performance tremendously!



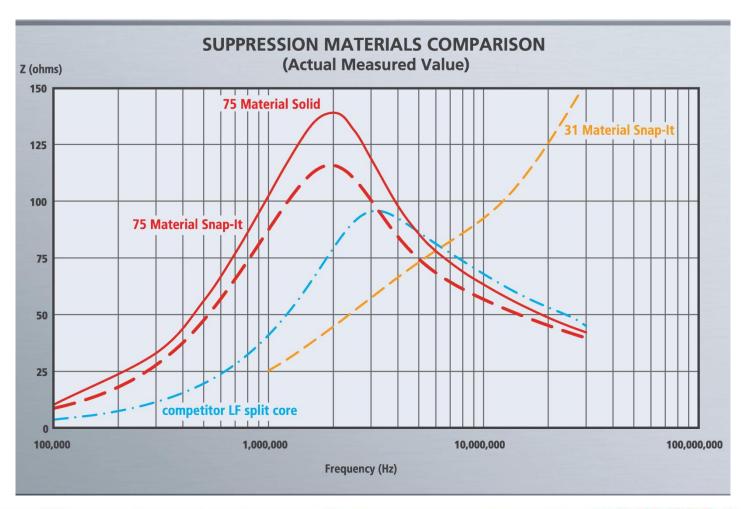


## Minimizing the air-gap





#### The solution!





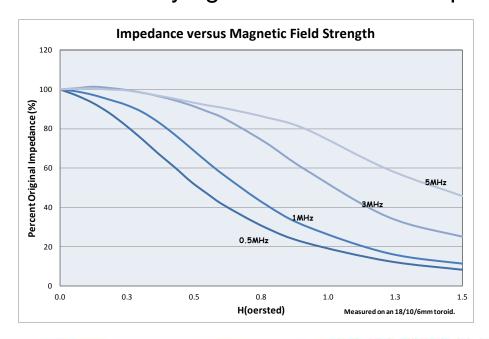
## Design Considerations

#### DC Bias

 Derating can occur at high current levels, so ferrites are ideal for common-mode noise issues.

When applicable, both current-carrying conductors should pass

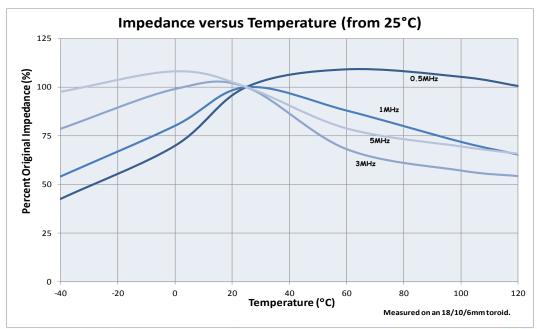
through the core.





## Design Considerations

- Operating Temperature
  - μ<sub>i</sub> for 75-material can change 0.6% per degree Celsius in an operating range from 20° 70°C.
  - μ<sub>i</sub> for 31-material can change 1.6% per degree Celsius in an operating range from 20° 70°C.





### Thank You!

• Questions?



